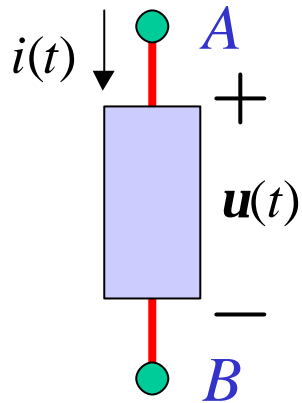


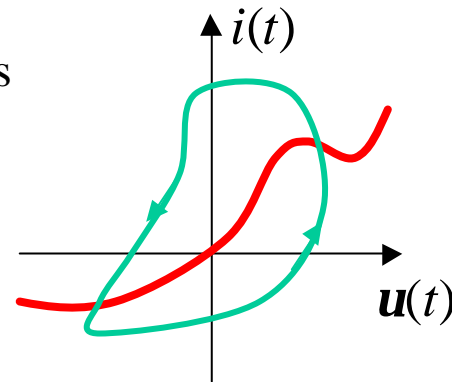
0.4 單埠元件與模型

單埠(one-port)

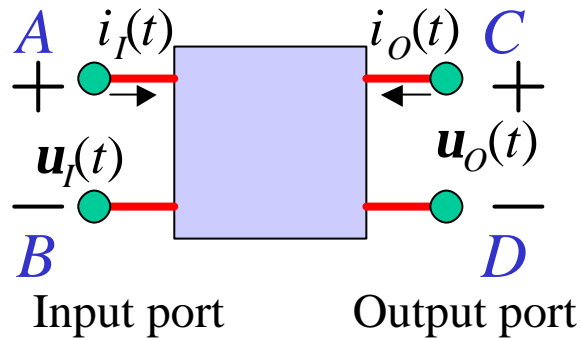


知道 $i(t) = i[u(t)]$ 或 $u(t) = u[i(t)]$ 即能瞭解此元件在電路中之行為及應用

- Voltage-ampere characteristics
- Voltage-ampere curve
- Model
- Constraint



雙埠(two-port)



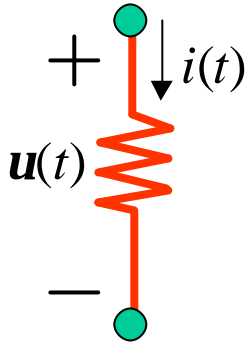
知道 (i_I, u_I) 與 (i_O, u_O) 之關係即能瞭解此元件在電路中之行為及應用

Two constraints

一些理想的單埠電路模型

電阻

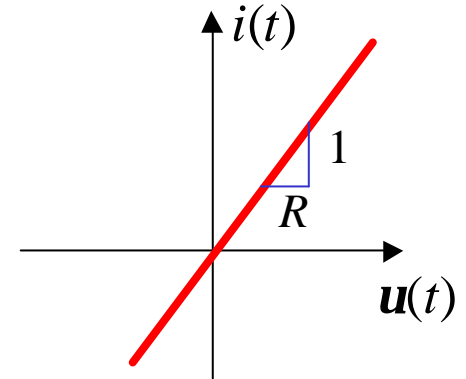
合乎Ohm's law的元件



$$i(t) = \left(\frac{1}{R} \right) \mathbf{u}(t)$$
$$\mathbf{I}(\mathbf{w}) = \left(\frac{1}{R} \right) \mathbf{V}(\mathbf{w})$$

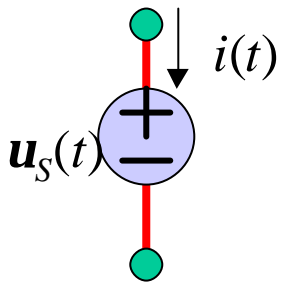
Time domain

Frequency domain



獨立電壓源

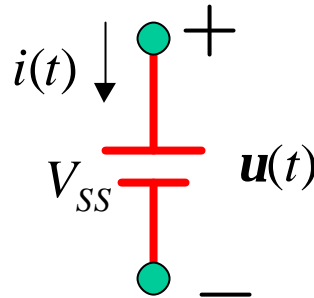
Independent voltage source



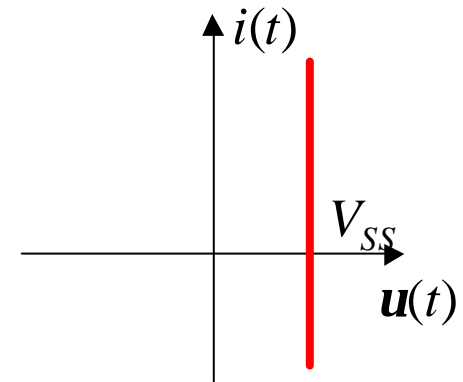
$$\mathbf{u}(t) = \mathbf{u}_s(t)$$
$$i(t) = ?$$

(電流由外界
電路決定)

直流獨立電壓源

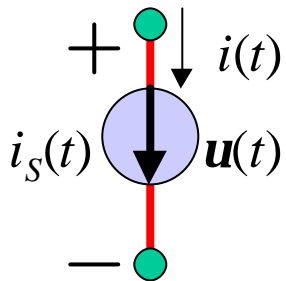


$$\mathbf{u}(t) = V_{SS}$$
$$i(t) = ?$$



獨立電流源

Independent current source

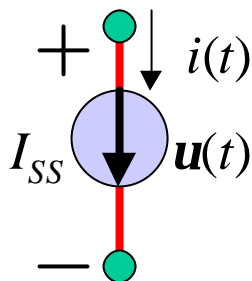


$$i(t) = i_s(t)$$

$$u(t) = ?$$

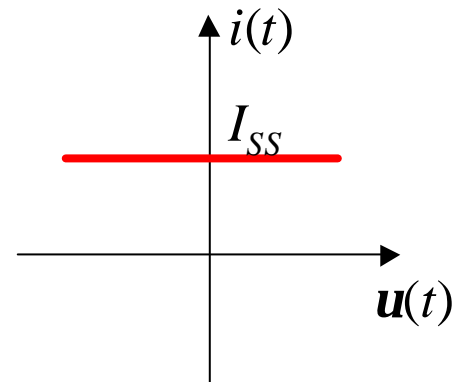
(電壓由外界
電路決定)

直流獨立電流源

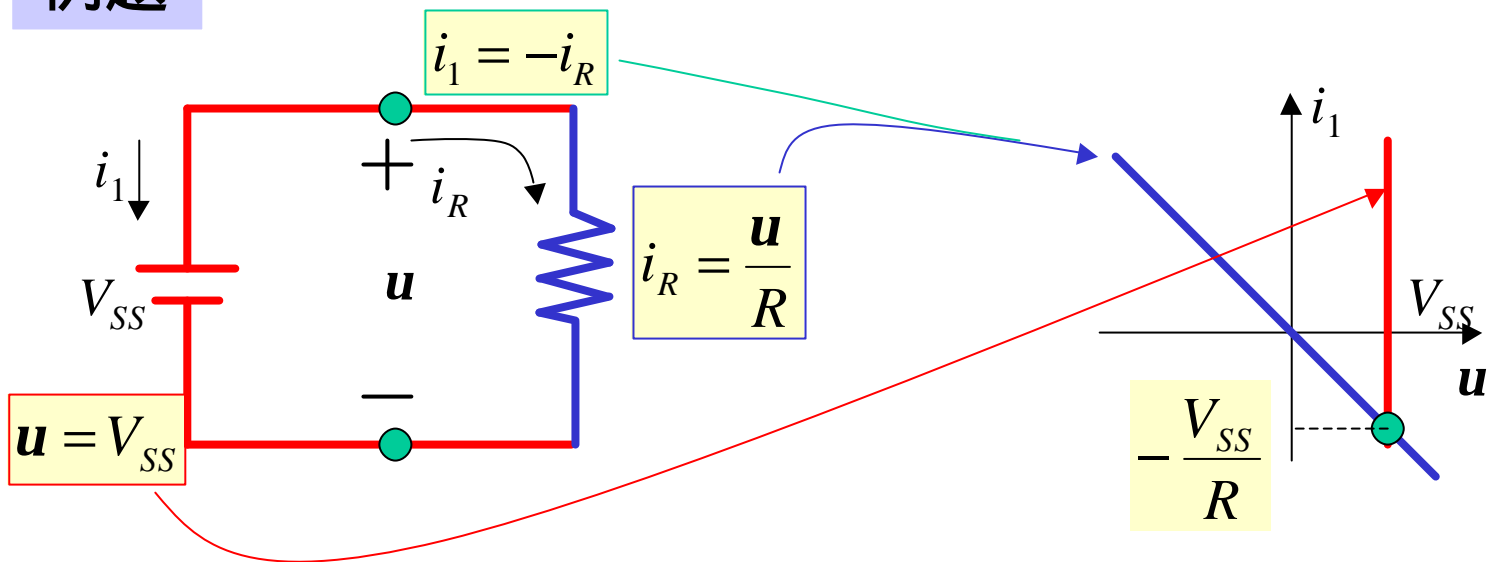


$$i(t) = I_{SS}$$

$$u(t) = ?$$

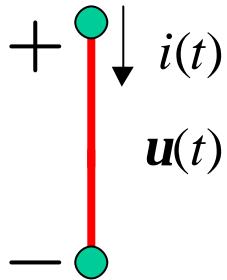


例題



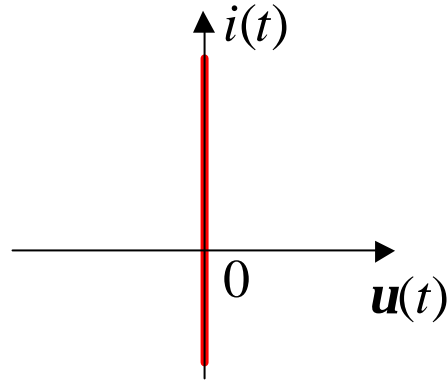
短路

Short circuit



$$\begin{aligned} u(t) &= 0 \\ i(t) &= ? \end{aligned}$$

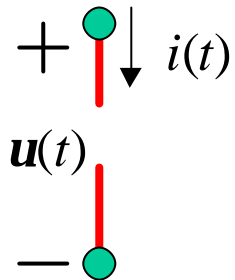
(電流由外界
電路決定)



- 為電壓源之特例
- 當電壓源關掉時，即成為短路

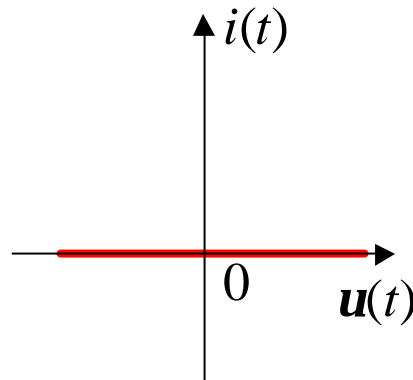
開路

(斷路) Open circuit



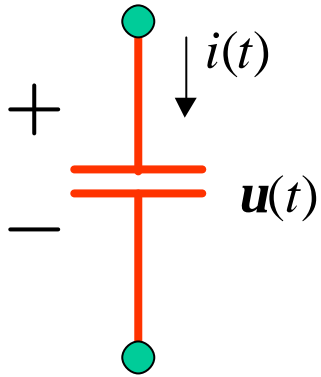
$$\begin{aligned} i(t) &= 0 \\ u(t) &= ? \end{aligned}$$

(電壓由外界
電路決定)



- 為電流源之特例
- 當電流源關掉時，即成為開路

電容



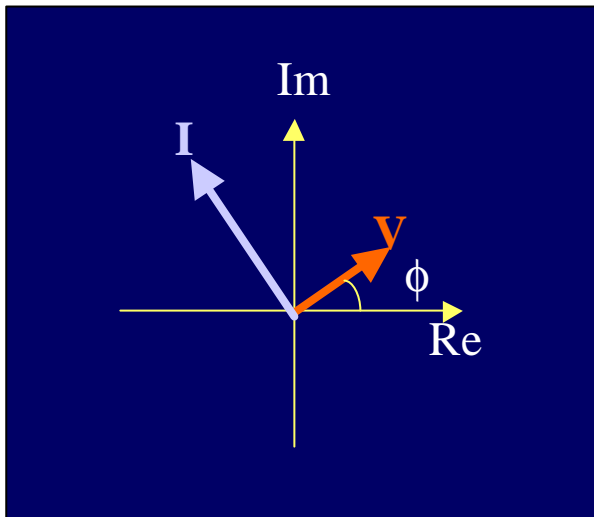
$$i(t) = C \frac{du(t)}{dt} \quad u(t) = V_0 \cos(\omega t + \mathbf{f}) \rightarrow \mathbf{V} = V_0 e^{j\mathbf{f}}$$

$$i(t) = -\omega C V_0 \sin(\omega t + \mathbf{f}) \rightarrow \mathbf{I} = j\omega C V_0 e^{j\mathbf{f}} = j\omega C \mathbf{V}$$

$$i(t) = \text{Re} \left[e^{j\mathbf{p}} \omega C V_0 e^{j\mathbf{f}} e^{j\omega t} e^{-\frac{j\mathbf{p}}{2}} \right]$$

$$\mathbf{V} = \mathbf{I} \mathbf{Z}_C$$

$$\mathbf{Z}_C = \frac{\mathbf{V}}{\mathbf{I}} = \frac{1}{j\omega C}$$



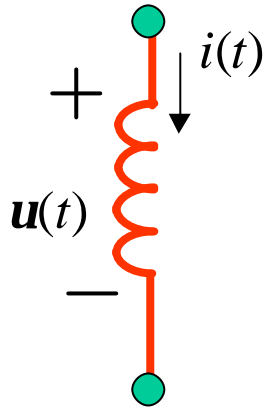
$$\mathbf{Z}_C = \frac{1}{j\omega C}$$

impedance

$$X_C \equiv \frac{1}{\omega C}$$

reactance

電感



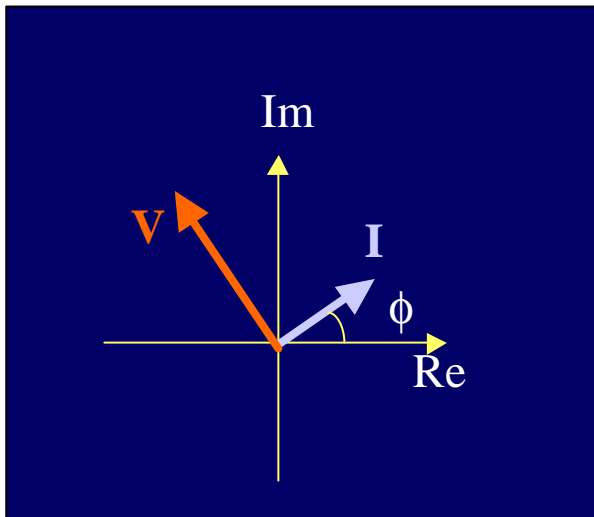
$$u(t) = L \frac{di(t)}{dt} \quad i(t) = I_0 \cos(\omega t + \mathbf{f}) \rightarrow \mathbf{I} = I_0 e^{j\mathbf{f}}$$

$$u(t) = -\omega L I_0 \sin(\omega t + \mathbf{f}) \rightarrow \mathbf{V} = j\omega L I_0 e^{j\mathbf{f}} = j\omega L \mathbf{I}$$

$$u(t) = \text{Re} \left[e^{j\mathbf{p}} \omega L I_0 e^{j\mathbf{f}} e^{j\omega t} e^{-\frac{j\mathbf{p}}{2}} \right]$$

$$\mathbf{V} = \mathbf{I} \mathbf{Z}_L$$

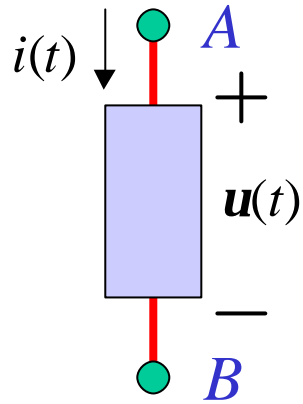
$$\mathbf{Z}_L = \frac{\mathbf{V}}{\mathbf{I}} = j\omega L$$



$$\mathbf{Z}_L = j\omega L$$

$$X_L = \omega L$$

單埠電路所消耗的功率



$$P(t) = i(t)u(t)$$

瞬時功率(instantaneous power)

$$P_{\text{av}}(t) = \lim_{t \rightarrow \infty} \frac{1}{t} \int_0^t i(t)u(t)dt$$

平均功率(average power)

考慮一AC電流通過某阻抗為 Z 的元件

$$\mathbf{Z} = |Z| e^{jf}$$

$$i(t) = I_{\text{max}} \cos \omega t$$

$$u(t) = V_{\text{max}} \cos(\omega t + f)$$

$$P(t) = i(t)u(t) = I_{\text{max}} V_{\text{max}} \cos \omega t \cos(\omega t + f)$$

$$P_{\text{av}}(t) = \lim_{t \rightarrow \infty} \frac{1}{t} \int_0^t i(t) \mathbf{u}(t) dt$$

$$= \frac{1}{T} \int_0^T i(t) \mathbf{u}(t) dt \quad \text{for periodic signal with period of } T$$

$$P_{\text{av}} = \frac{1}{T} \int_0^T I_{\text{max}} V_{\text{max}} \cos \omega t \cos(\omega t + \mathbf{f}) dt$$

$$= \frac{I_{\text{max}} V_{\text{max}}}{T} \int_0^T \cos \omega t [\cos \omega t \cos \mathbf{f} - \sin \omega t \sin \mathbf{f}] dt$$

$$= \frac{I_{\text{max}} V_{\text{max}}}{T} \int_0^T [\cos^2 \omega t \cos \mathbf{f} - \cos \omega t \sin \omega t \sin \mathbf{f}] dt$$

$$= \frac{I_{\text{max}} V_{\text{max}}}{2} \cos \mathbf{f}$$

$$V_{\text{max}} = I_{\text{max}} |Z|$$

$$P_{\text{av}} = \frac{I_{\text{max}} V_{\text{max}}}{2} \cos \mathbf{f} = \frac{I_{\text{max}}^2 |Z|}{2} \cos \mathbf{f} = \frac{V_{\text{max}}^2}{2 |Z|} \cos \mathbf{f}$$

我們稱 $\cos \phi$ 為功率因子(power factor)， ϕ 恰為元件或電路阻抗的相角。

對於純電容或電感的元件， ϕ 為 $\pi/2$ ， $\cos\phi=0$ ，平均消耗功率為零。電容與電感為能量儲存元件，不消耗能量。電阻則為能量消耗元件， $\cos\phi=1$ 。

對電阻而言

$$P_{avR} = \frac{I_{\max} V_{\max}}{2} = \frac{I_{\max}^2 R}{2} = \frac{V_{\max}^2}{2R}$$

我們可以定義一方便的平均值使得

$$I_{av}^2 = \frac{I_{\max}^2}{2} \quad V_{av}^2 = \frac{V_{\max}^2}{2}$$

即
$$I_{av} = \frac{I_{\max}}{\sqrt{2}} = 0.707I_{\max} \quad V_{av} = \frac{V_{\max}}{\sqrt{2}} = 0.707V_{\max}$$

這個平均恰好是所謂的方均根(rms)值

$$\begin{aligned} P_{av} &= I_{av} V_{av} \cos f = I_{av}^2 |Z| \cos f = \frac{V_{av}^2}{|Z|} \cos f \\ &= I_{rms} V_{rms} \cos f = I_{rms}^2 |Z| \cos f = \frac{V_{rms}^2}{|Z|} \cos f \end{aligned}$$

或用複數表示

$$\begin{aligned} P_{av} &= \text{Re}[\mathbf{I}\mathbf{V}^*] = \text{Re}[\mathbf{I}^* \mathbf{V}] \\ &= I_{rms} V_{rms} \cos f \end{aligned}$$

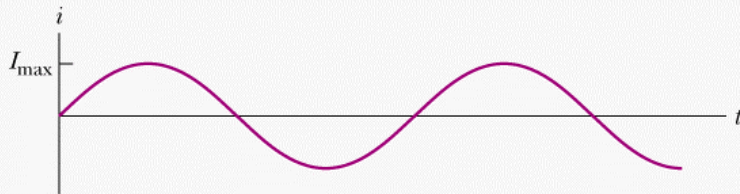
方均根值(root mean square, rms)

他的基本定義：

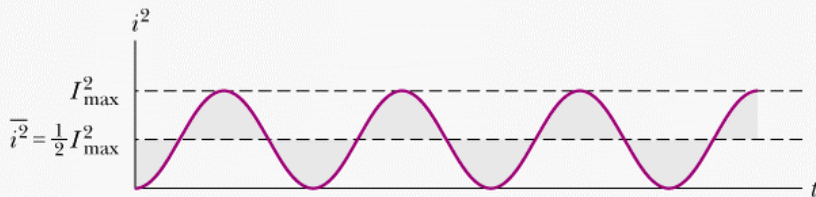
$$f_{\text{rms}} = \sqrt{\frac{1}{T} \int_0^T [f(t)]^2 dt}$$

假如 $f(t)$ 是平均為0的弦波函數，即 $f(t) = f_{\text{max}} \cos(\omega t + \mathbf{q})$

$$f_{\text{rms}} = \sqrt{\frac{1}{T} \int_0^T [f_{\text{max}} \cos(\omega t + \mathbf{q})]^2 dt} = \frac{|f_{\text{max}}|}{\sqrt{2}}$$



(a)



(b)

假如訊號不是弦波或平均值不是0，則上式便不成立。